Multiple Description Coding Over Many Channels

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Outline

- Motivation for *Multiple Description Coding*
- Problem statement
- Review of known results (Two-Description Coding)
- New results for L-description coding $(L > 2)$
- Special cases and examples
- Summary

Underwater Wireless Channels

- Limited bandwidth and severe multipath propagation.
- Long delays (speed of sound in water is 1500m/s)
- Need for transmission of real-time data:
	- e.g., images or video data to/from divers.
- Poor propagation occasionally cause weak SNRs.

• Packets are either lost completely (erased) or received error-free

How do we deal with erasures?

- Request a retransmission.
	- Ideal for loss-less transmission.
	- Not feasible for real time data such as voice and video.

Alternate Approach:

- Reconstruct using available packets
	- Requires adding redundancy to packets, i.e., coding across packets.

This approach is called Multiple Description Coding.

Transmitting Independent Packets

Transmitting Correlated Packets

Multiple Description (MD) Coding vs. Compression

- Compression (no coding across packets):
	- Best reconstruction when all packets are received.
	- Sharp degradation as packet losses increase.
- Multiple Description Coding:
	- Graceful degradation as packet losses increase.
	- Suboptimal when all packets are received.

Source: Length N vector X^N of i.i.d. random variables.

Encoder: $X^N \to \{J_1, \ldots, J_L\}$ which are the L "descriptions" of X^N at rates R_1, \ldots, R_L per source symbol.

Descriptions:

 $J_l = f_l(X^N), \quad H(J_l) \le NR_l, \quad l = 1, \ldots L.$

Decoder: Consists of $2^L - 1$ sub-decoders: one for each non-empty subset of the available descriptions.

Decoder Outputs: $X_{\mathcal{S}}^{N} = g_{\mathcal{S}}(\{J_{l}:l\in\mathcal{S}\})$ where $\mathcal{S} \subseteq \{1, \ldots, L\}, \mathcal{S} \neq \emptyset.$

Example: Coding with 3 Descriptions

- The source is i.i.d. vector X^N (length N).
- The encoder outputs L descriptions J_1, J_2, J_3 of X^N .
- \bullet The decoder produces an output $\hat X^N$ from the available descriptions.

Problem Statement

- Problem: What is the Rate-Distortion (R-D) region?
- The rates (L parameters) are R_1, \ldots, R_L .
- The distortions $((2^L-1)$ parameters) are

$$
D_{\mathcal{S}} = \frac{1}{N} \mathrm{E} d(X^N, X^N_{\mathcal{S}}), \quad \mathcal{S} \subseteq \{1, \ldots, L\}, \mathcal{S} \neq \emptyset
$$

• The R-D region is $(L + 2^L - 1)$ -dimensional.

- Why do we need the R-D region?
	- Provides complete knowledge about tradeoff between packet redundancy and fidelity of reconstruction.
	- Helps design good MD coders.
- For $L = 1$, it is Shannon's R-D region.

Rate-Distortion Theory

The R-D region is the set of achievable (R, D) as $N \to \infty$.

Theorem 1. [Shannon] The R-D region is the convex set $R \geq R(D)$ where

$$
R(D) = \min_{\hat{X}} I(X; \hat{X}) \quad \text{s.t.} \quad \mathrm{E}d(X, \hat{X}) \le D
$$

minimized over all \hat{X} jointly distributed with X .

For the Gaussian Source: $X \sim N(0, 1)$, the rate-distortion fucntion is

$$
R(D) = \frac{1}{2} \log \left(\frac{1}{D}\right)
$$

Two-Description Coding

The rates and distortions are:

$$
R_1 = \frac{1}{N}H(J_1)
$$

\n
$$
B_2 = \frac{1}{N}H(J_1)
$$

\n
$$
D_1 = \frac{1}{N}Ed(X^N, X_1^N)
$$

\n
$$
D_2 = \frac{1}{N}Ed(X^N, X_2^N)
$$

\n
$$
D_{12} = \frac{1}{N}Ed(X^N, X_{12}^N)
$$

Two-Description Coding

El Gamal and Cover (1982) found an achievable region for $L = 2$:

$$
R_1 \ge I(X; X_1)
$$

\n
$$
R_2 \ge I(X; X_2)
$$

\n
$$
R_1 + R_2 \ge I(X; X_1X_2X_{12}) + I(X_1; X_2)
$$

\n
$$
D_S \ge Ed(X, X_S), \quad S = 1, 2, 12
$$

where X_1 , X_2 , X_{12} are any r.v's jointly distributed with the source X .

- The convex hull of this region is achievable by time-sharing.
- This is an achievable region, not necessarily the $R-D$ region.
- Ozarow (1980) computed the R-D region for the Gaussian Source.

The Gaussian Source

The Gaussian rate region

- Ahlswede (1987) showed tightness of the inner bound by El Gamal and Cover in the "no excess rate" case $(R_1 + R_2 = R(D_{12})$ for the joint description output: X_{12} .
- Zhang and Berger (1987) provided a stronger achievable result than El Gamal and Cover for $L=2$. For the **binary symmetric source** with Hamming distortion measure, their result provides a strict improvement.

Progressive Data Transmission

The decoding is **progressive**:

$$
(J_1) \to D_1 \ge D(R_1)
$$

$$
(J_1, J_2) \to D_{12} \ge D(R_1 + R_2)
$$

$$
(J_1, J_2, J_3) \to D_{123} \ge D(R_1 + R_2 + R_3)
$$

- Question: Can we attain all the minimum distortions simultaneously?
- Answer: Not always. X is successively refinable if the Shannon bounds are achievable with equality:

$$
D_1 = D(R_1)
$$

\n
$$
D_{12} = D(R_1 + R_2)
$$

\n
$$
D_{123} = D(R_1 + R_2 + R_3)
$$

Successive Refinement

• The Gaussian source is successively refinable:

R-D region: $R_1 \ge R(D_1)$, $R_1 + R_2 \ge R(D_{12})$.

• This structure is a special case of MD coding with X_2 absent.

Summary of New Results

- Achievable region for MD coding of an arbitrary source $(L \geq 2).$
- R-D region bounds for the Gaussian source.
- Special cases of Gaussian MD coding.
- Examples.

Achievable Region for $L \geq 2$

Theorem 2. The R-D region contains the rates and distortions satisfying

$$
\sum_{l \in S} R_l \ge (|S| - 1)I(X; X_{\emptyset}) - H(X_{\mathcal{U}} : \mathcal{U} \in 2^S | X)
$$

$$
+ \sum_{\mathcal{T} \subseteq S} H(X_{\mathcal{T}} | X_{\mathcal{U}} : \mathcal{U} \in 2^{\mathcal{T}} - \mathcal{T})
$$

$$
D_S \ge Ed_{\mathcal{S}}(X, X_{\mathcal{S}})
$$

for every $\emptyset \neq \mathcal{S} \subseteq \mathcal{L} = \{1, \ldots, L\}$ and some joint distribution between outputs $\{X_{\mathcal{S}}\}$ and the source X .

- Best region is obtained by optimizing over the joint **p.d.f** of $(X, \{X_{\mathcal{S}} : \mathcal{S} \in 2^{\mathcal{L}}\})$.
- For a given p.d.f, the rate-region is a polyhedron:
	- Every constraint is active (polymatroid structure).
- Region is an inner bound on the R-D region.
- Proof uses random coding.
- For $L = 2$, this produces the results of **El Gamal and** Cover, and of Zhang and Berger.

Outer Bound on the R-D Region

- Outer bounds: difficult to find for arbitrary sources.
	- Proof depends strongly on properties of source.
- We focus on
	- Gaussian source: $X \sim N(0, 1)$.
	- Squared-error distortion: $d(x,y) = |x-y|^2$.

• An outer bound on the R-D region: **Theorem 3.** The R-D region is contained in

$$
\sum_{k \in \mathcal{K}} R_k \ge -\frac{1}{2} \log \left[\min_{\{\mathcal{K}_m\}_{m=1}^M} \inf_{\lambda \ge 0} \left(D_{\mathcal{K}} \frac{\prod_{m=1}^M (D_{\mathcal{K}_m} + \lambda)}{(D_{\mathcal{K}} + \lambda)(1 + \lambda)^{M-1}} \right) \right], \quad \forall \mathcal{K} \in 2^{\mathcal{L}}
$$

minimized over all partitions $\{K_m\}$ of K.

- The outer bound is a polyhedron.
- We also have a stronger outer bound.

L Channels and $L+1$ Decoders

Keep only **side** and **central** decoders. Ignore all other decoders.

Bounds on the R-D region

• Outer Bound: For the L-channel $(L + 1)$ -receiver problem, the outer bound reduces to

$$
R_l \ge -\frac{1}{2}\log D_l, \quad (l = 1, \dots, L)
$$

$$
R_1 + \dots + R_L \ge -\frac{1}{2}\log \varphi_L(D_0, \dots, D_L)
$$

$$
\varphi_L(D_0, \dots, D_L) = \inf_{\lambda \ge 0} \left[D_0 \frac{\prod (D_l + \lambda)}{(D_0 + \lambda)(1 + \lambda)^{L-1}} \right]
$$

Inner Bound

Pick a joint Gaussian p.d.f. for $(X, \{X_{\mathcal{S}} : \mathcal{S} \in 2^{\mathcal{L}}\})$: Let (W_1, \ldots, W_L) be jointly Gaussian with covariance matrix K .

$$
X_l = E(X|X + W_l) \equiv \alpha_l(X + W_l), \quad (l = 1, \dots, L)
$$

$$
X_0 = E(X|X_1, \dots, X_L) \equiv \sum_{l=1}^L \beta_l X_l
$$

for appropriate α_l , β_l .

- For given distortions D_1, \ldots, D_L and D_0 , we optimize over K to find the best rate region.
- The optimal solution takes the form:

$$
\bm{K} = \left(\begin{matrix} \sigma_1^2 & u & u & \dots & u \\ u & \sigma_2^2 & u & \dots & u \\ u & u & \sigma_3^2 & \dots & u \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ u & u & u & \dots & \sigma_L^2 \end{matrix}\right)
$$

Tightness of Bounds

The inner and outer bounds meet for some rates and distortions.

Theorem 4. For any $L \geq 2$ and distortions D_0, D_1, \ldots, D_L , the best achievable sum rate meets the outer bound:

$$
\sum_{l=1}^L R_l = 1\frac{1}{2}\log \varphi_L(D_0,\ldots,D_L).
$$

Example: 3-Channel 4-Decoder Problem

Take
$$
L = 3
$$
, $D_1 = D_2 = D_3 = 1/2$ and $D_0 = 1/16$.

Outer Bound:

$$
R_l \ge 0.5, \quad l = 1, 2, 3
$$

$$
R_1 + R_2 + R_3 \ge 2.1755
$$

Achievable Rates:

$$
R_l \ge 0.5, \quad l = 1, 2, 3
$$

$$
R_1 + R_2 + R_3 = 2.1755
$$

$$
R_l + R_m \ge 1.1258, \quad l < m
$$

Blue Region: Inner and outer bounds meet on a hexagon. Green Region: Inner bound (does not meet outer bound).

Theorem 5. Suppose

$$
D_L + \Big(1-L+\sum_{l=1}^{L-1} D_l^{-1}\Big)^{-1} \geq 1+D_0.
$$

then the outer and inner bounds meet everywhere.

Example:
$$
L = 3
$$
, $D_1 = D_2 = 1/2$, $D_3 = 3/4$, and $D_0 = 1/16$:

The above inequality

$$
D_3 + \frac{D_1 D_2}{D_1 + D_2 - D_1 D_2} \ge 1 + D_0
$$

is satisfied.

The R-D Region

Successive Refinement on Chains

$$
X \xrightarrow{R_1} X_1 \xrightarrow{R_2} X_{12} \xrightarrow{R_3} X_{123} \dots
$$

Gaussian

 $D_1 \ge D(R_1)$ $D_{12} \ge D(R_1 + R_2)$ $D_{123} \geq D(R_1 + R_2 + R_3)$

Proved using the achievability result.

Successive Refinement on Trees

 $D_1 \geq D(R_1)$ $D_{12} \ge D(R_1 + R_2)$ $D_{123} \geq D(R_1 + R_2 + R_3)$ $D_{124} \geq D(R_1 + R_2 + R_4)$ $D_{15} \ge D(R_1 + R_5)$

Summary

- Multiple Description Coding:
	- Discussed applications to real-time transmission over underwater acoustic networks.
	- Reviewed known MD coding results.
- Our new results:
	- Achievable rate region for MDC with *many* channels.
	- Outer bound on the R-D region for the Gaussian source.
	- Demonstrated tightness of bounds.